

The seal of the University of Gothenburg is a circular emblem. It features a central figure, a lion rampant, holding a sword aloft in its right paw and resting its left paw on an open book. The lion is crowned with a crown. The text around the inner border of the seal reads "SIGILLVM UNIVERSITATIS GOTHORVM CAROLINAE VT RVMQVE" and the year "1666" is inscribed at the bottom.

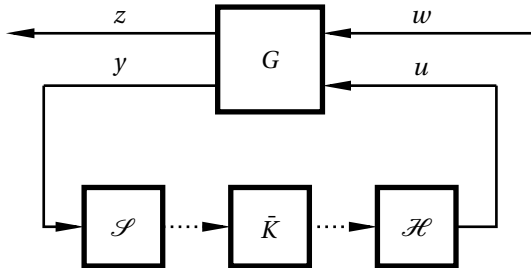
On Optimal Sampling for Event-Based Control

Marcus T. Andrén



Problem

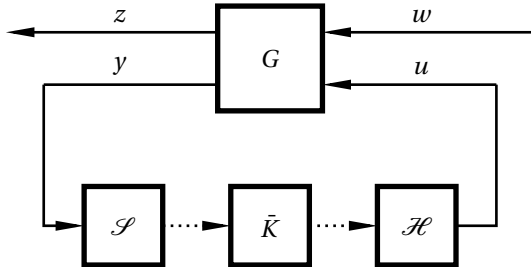
How to optimally co-design controller structure and sampling strategy?





Problem

How to optimally co-design controller structure and sampling strategy?



Goal

Minimize

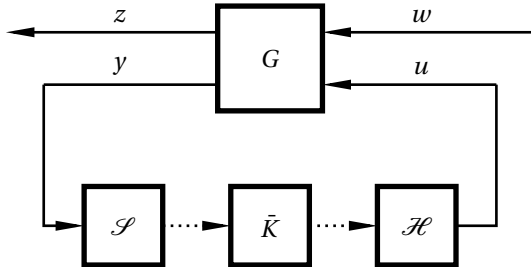
$$\|T_{zw}\|_2^2 := \lim_{T \rightarrow \infty} E\left[\frac{1}{T} \int_0^T \|z(t)\|_2^2 dt\right]$$

w.r.t $\mathcal{S}, \bar{K}, \mathcal{H}$ and sampling



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How to optimally co-design controller structure and sampling strategy?



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Why?

- Most efficient use of sampling
- To lower energy consumption
- To reduce risk for congestion



Lower Bound on $\|T_{zu}\|_2^2$

$$G: \begin{bmatrix} \dot{x} \\ z \\ y \end{bmatrix} = \left[\begin{array}{c|cc} A & B_w & B_u \\ \hline C_z & 0 & D_{zu} \\ C_y & D_{yw} & 0 \end{array} \right] \begin{bmatrix} x \\ w \\ u \end{bmatrix}$$

Without restrictions on sampling, analog LQG is optimal:

LQR Gain:

$$L = (D_{zu}^T D_{zu})^{-1} (B_u^T X + D_{zu}^T C_z)$$

Kalman Gain:

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with optimal cost

$$\|T_{zw}\|_2^2 = \gamma_0^2 = \text{Tr}(B_w^T X B_w) + \text{Tr}(C_z Y C_z^T) + \text{Tr}(X A Y + Y A^T X)$$

- No sampled-data controller can beat γ_0^2
- How close can we get?



The Optimal Controller Structure

[Mirkin 2017, Th 5.1]

The optimal attainable \mathcal{H}_2 performance using sampled-data controllers for any sampling sequence $\{t_i\}$ is

$$\|T_{zw}\|_2^2 = \gamma_0^2 + \|H\|_2^2$$



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where H is a reset system driven by white process ϵ with intensity $D_{yw}D_{yw}^T$:

$$H: \begin{cases} \dot{x}_H(t) = Ax_H(t) - K\epsilon, & x_H(t_i) = 0 \\ \eta(t) = -(D_{zu}^T D_{zu})^{\frac{1}{2}} Lx_H(t) \end{cases}$$

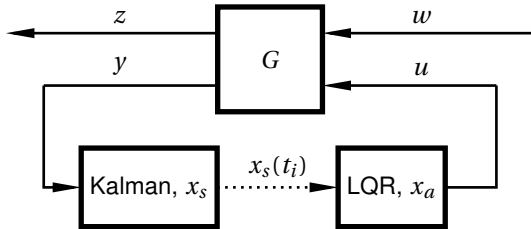
Degradation from sampling given by average variance of η !



The Optimal Sampled-Data Controller

[Mirkin 2017]

The optimal sampled-data controller can be realized as:



Sensor-side

$$\dot{x}_s(t) = Ax_s(t) + B_u u(t) + K(y(t) - C_y x_s(t))$$

Remark: $x_H = x_a - x_s$ and $\epsilon = y - C_y x_s$

Actuator-side

$$\dot{x}_a(t) = (A - B_u L)x_a(t)$$

$$x_a(t_i) = x_s(t_i)$$

$$u(t) = -Lx_a(t)$$



Optimal Sampling?

Structure in [Mirkin 2017] optimal for *any uniformly bounded* $\{t_i\}$
 \Rightarrow Sampling scheme can be considered separately!

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- No sampling cost \implies analog LQG. "Sample infinitely often"
- Per-sample-cost $\rho \implies$ A new optimization problem.



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Define $\phi(x_H) : x_H \rightarrow \{0, 1\}$ and let $\{t_i\} = \{t | \phi(x_H(t)) = 1\}$.

Optimal Sampling Goal:

$$\min_{\phi} \underbrace{\|H\|_2^2}_{\text{Performance Degradation}} + \underbrace{\lim_{T \rightarrow \infty} \frac{1}{T} E[\sum_i \phi(x_H(t_i))]}_{\text{Average Sampling Rate}} \cdot \underbrace{\rho}_{\text{Per-sample-cost}}$$



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Reset System:

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$$\eta(t) = x(t)$$

where ϵ is a white process with intensity R .



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Cost:

$$J = \lim_{T \rightarrow \infty} \frac{1}{T} E \left[\int_0^T x^T Q x dt + \rho \{ \# \text{ events up to } T \} \right]$$



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Note: Identical to our optimization problem if:

$$Q = L^T D_{zu}^T D_{zu} L \quad R = K D_{yw} D_{yw}^T K^T$$



Solution?

Optimal sampling problem can be re-formulated as finding a *value function* $V(x)$:

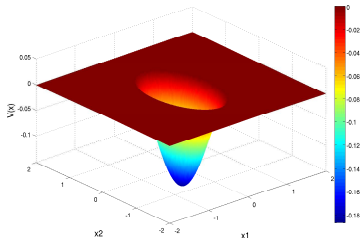
[Henningsson 2012, Paper II, Th 1]

Suppose a bounded function $V(x)$ and constant J are found satisfying

$$x^T Q x + x^T A^T \nabla V + \frac{1}{2} \text{Tr}(R \nabla^2 V) \geq J \quad (1)$$

$$\rho \geq V(x) - V(0) \quad (2)$$

where for each x equality is achieved in either (1) or (2). Then the optimal cost is J and it's optimal to trigger when equality is achieved in (2).





Integrator Case: $A = 0$

[Henningsson 2012]

A closed-form solution exists for the integrator case:

$$V(x) = \begin{cases} -\frac{1}{4}g(x)^2, & g(x) \geq 0 \\ 0 & \text{else} \end{cases}$$

where

$$g(x) = 2\sqrt{\rho} - x^T P x$$

and P is the solution to the Riccati-like equation

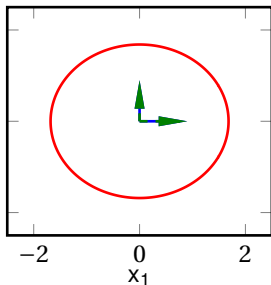
$$PRP + \frac{1}{2} \text{Tr}(RP)P = Q$$



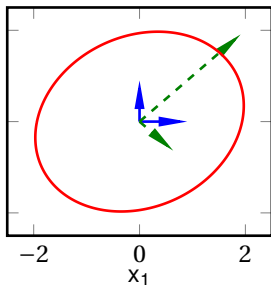
Integrator Case: $A = 0$

The optimal trigger bound is the ellipsoid $x^T P x = 2\sqrt{\rho}$.

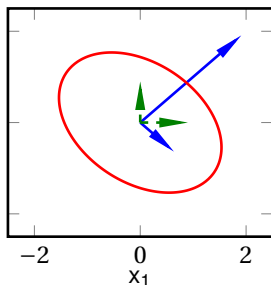
"Shape" of P depends on Q and R (here, $\rho = 1$):



$$Q = I$$
$$R = I$$



$$Q = I$$
$$R = \begin{bmatrix} 5.5 & 4.5 \\ 4.5 & 5.5 \end{bmatrix}$$



$$Q = \begin{bmatrix} 5.5 & 4.5 \\ 4.5 & 5.5 \end{bmatrix}$$
$$R = I$$



Non-optimal vs. Optimal Sampling?

Let's look at a 2nd order integrator example:

$$\left[\begin{array}{c|cc} A & B_w & B_u \\ \hline C_z & 0 & D_{zu} \\ C_y & D_{yw} & 0 \end{array} \right] = \left[\begin{array}{c|cc} 0 & [(R_d^{1/2} N_{\pi/8})^T & 0] & I \\ \hline [N_{\pi/4} Q_d^{1/2}] & 0 & \begin{bmatrix} 0 \\ I \end{bmatrix} \\ I & [0 & I] & 0 \end{array} \right]$$

$$N_\theta = \begin{bmatrix} \cos(\theta) & -\sin(\theta) \\ \sin(\theta) & \cos(\theta) \end{bmatrix} \quad R_d = Q_d = \text{diag}([1, 5])$$



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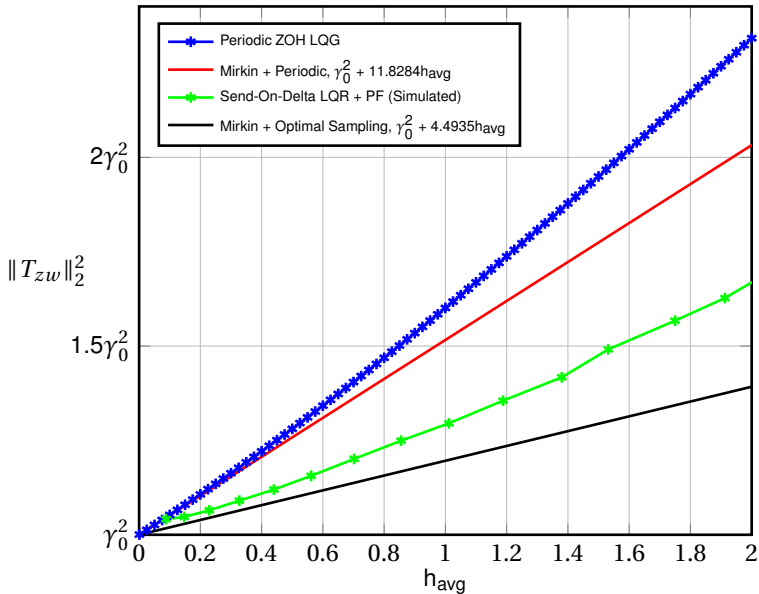
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- Periodic ZOH LQG
- [Mirkin 2017] + Periodic Sampling
- LQR with Particle Filter + Send-On-Delta Sampling
- [Mirkin 2017] + Optimal Sampling



Example, $A=0$





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Using optimal controller structure, optimal sampling was $\frac{11.8284}{4.4935} \approx 2.63$ times more efficient than periodic.



Example, A=0

Using optimal controller structure, optimal sampling was $\frac{11.8284}{4.4935} \approx 2.63$ times more efficient than periodic.

In fact, for any n^{th} order integrator system, the ratio is:

$$1 + \frac{2}{n} \leq \frac{\text{Periodic}}{\text{Optimal}} \leq 3$$



How About $A \neq 0$?

- The million \$ question (and current focus). Solving this \implies
A synthesis method for event-based control!

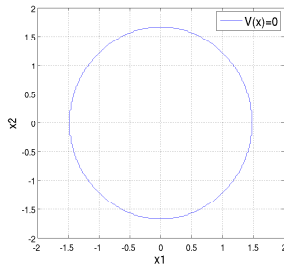
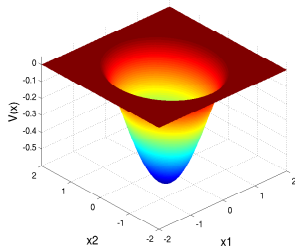


How About $A \neq 0$?

- The million \$ question (and current focus). Solving this \implies A synthesis method for event-based control!
- Currently we can obtain $V(x)$ for 2nd order systems numerically.
- Any hypothesis based on numerical results can be checked using PDE for $V(x)$.



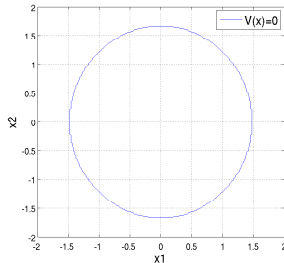
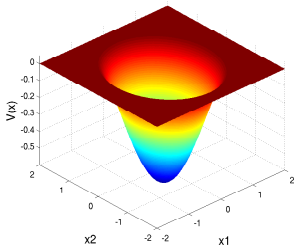
Numerically Obtained $V(x)$



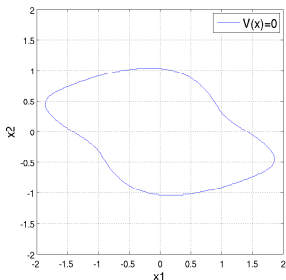
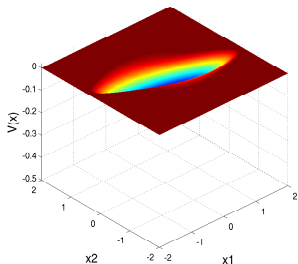
$$A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$$



Numerically Obtained $V(x)$



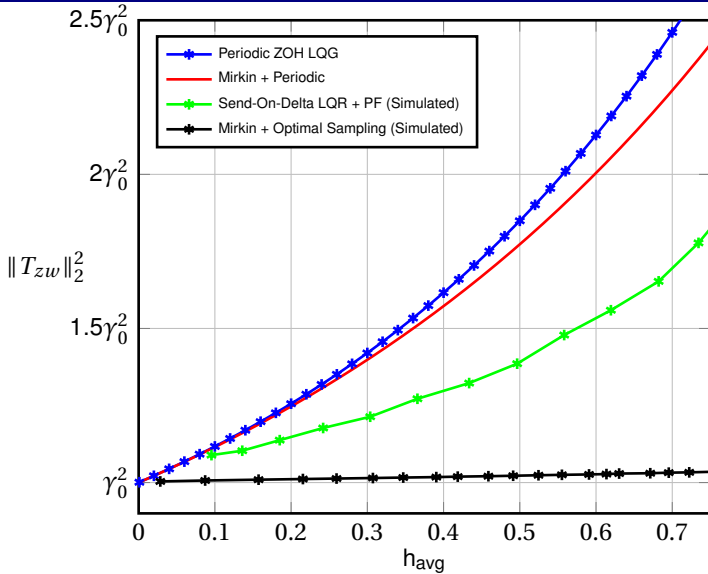
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$$A = \begin{bmatrix} 1 & -1 \\ 0 & -70 \end{bmatrix}$$



Example, $A = \begin{bmatrix} 1 & -1 \\ 0 & -70 \end{bmatrix}$





Conclusion

- Problem: Co-design of controller and sampling strategy
- Nice result from Mirkin makes the problem separable.
- Similar work by Toivo directly applicable to the sampling problem.
- $A = 0$ case solved, $A \neq 0$ only numerically.
- Any hypothesis can be verified/rejected with PDE for $V(x)$.

References:

- L. Mirkin, "Intermittent redesign of analog controllers via the Youla parameter," IEEE Trans. Automat. Control, vol.62, 2017, (to appear). [Online]. Available: <http://ieeexplore.ieee.org/stamp/stamp.jsp?arnumber=7523893>
- T. Henningson, "Stochastic event-based control and estimation," Ph.D. dissertation, Dept. of Automatic Control, Lund University, Lund, Sweden, 2012.